

Department of Electrical Engineering
University of Arkansas



ELEG 5693 Wireless Communications Math Review

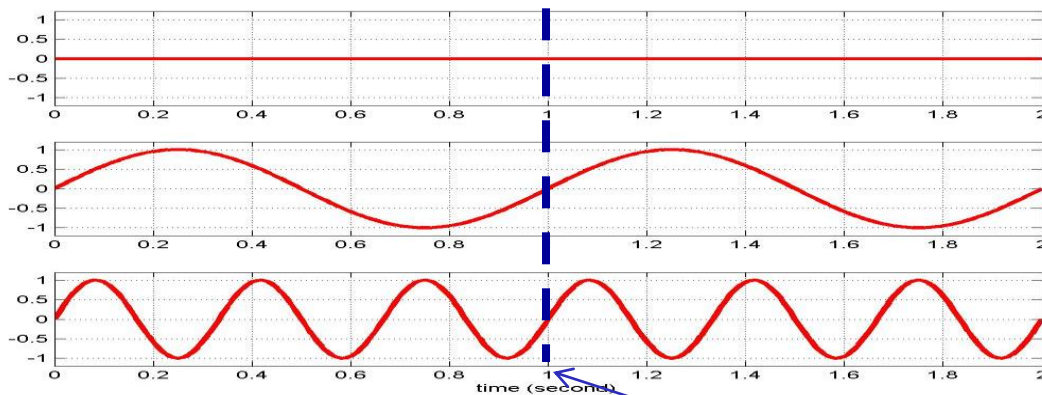
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OUTLINE

- **Signals and System (Fourier analysis)**
- **Random variables and random process**

CONCEPT OF FREQUENCY

- **Frequency is the measurement of the number of times that a repeated event occurred in a unit time (1 second).**
 - Frequency measures how fast a signal can change within a unit period of time (or the measurement of the rate of change).
 - High frequency → the signal changes fast
- **In signal processing, frequency is defined based on sinusoid signal → number of cycles per second.**
 - Each sinusoid signal is uniquely associated with a single frequency based on its rate of change.



$$f = 0 \text{ Hz}$$

$$f = 1 \text{ Hz}$$

$$f = 3 \text{ Hz}$$

1 second

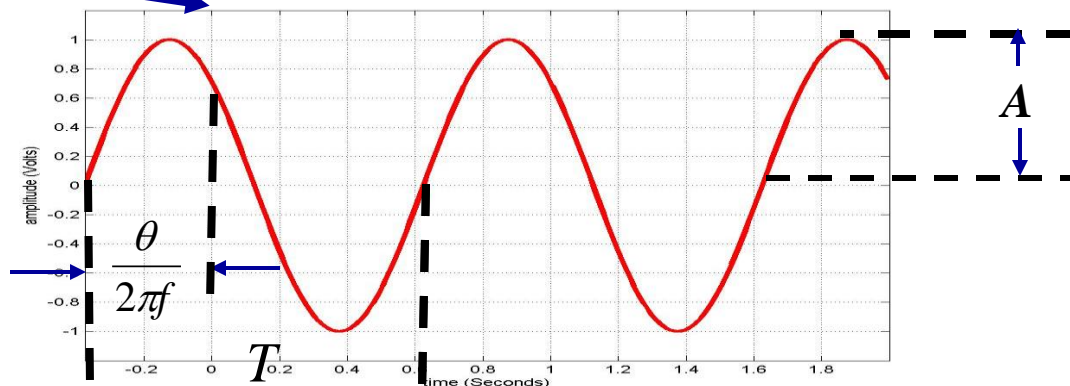
TIME DOMAIN SIGNAL

- **Sinusoid Signal**

$$s(t) = A \sin(2\pi ft + \theta)$$

- f : frequency. in the unit of Hz (1/second)
- A : amplitude. maximum strength in signal, usually in unit of volts
- θ : phase. relative position in time
- t : time. in the unit of second
- T : period. Time for one cycle or one repetition, $T=1/f$.
- λ : Wavelength. Distance of the signal waveform propagated in one period T .

0 second • $\lambda = vT$, where v is the speed of the signal. In vacuum, it is the speed of light.

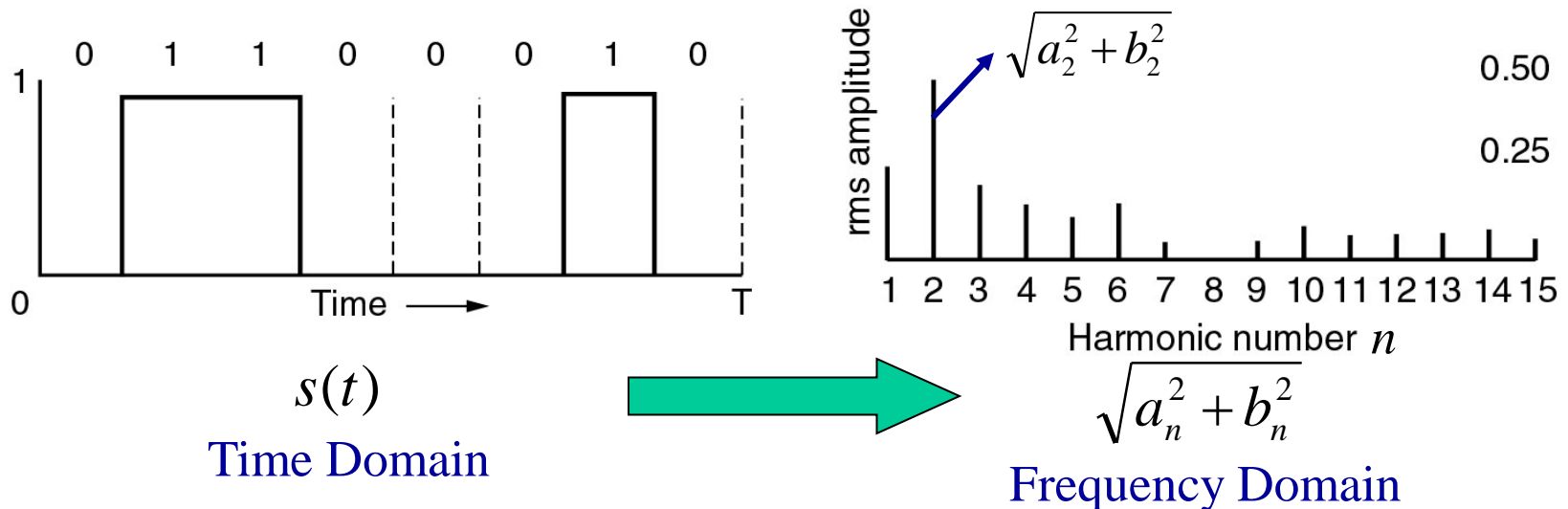


THEORY: FREQUENCY DOMAIN SIGNAL

- **Periodic signal can be written as the summation of a series of sinusoids (Fourier Series)**

$$s(t) = \frac{1}{2}c + \sum_{n=1}^{\infty} a_n \sin(2\pi n f_n t) + \sum_{n=1}^{\infty} b_n \cos(2\pi n f_n t)$$

- Each sinusoid component has a unique frequency
- Each sinusoid component is called a “harmonic” of the original signal.
- There is a one to one relationship between a signal and its Fourier series.
- With Fourier series, we can represent the signal in the “**Frequency Domain**”



FOURIER TRANSFORM

Let $g(t)$ be a non-periodic deterministic signal, expressed as function of t . Its frequency domain representation, $G(f)$, can be obtained from Fourier transform.

Fourier transform

$$G(f) = \int_{-\infty}^{+\infty} g(t) \exp(-j2\pi ft) dt$$

Inverse Fourier transform

$$g(t) = \int_{-\infty}^{+\infty} G(f) \exp(j2\pi ft) dt$$

$$j^2 = -1 \qquad \exp(jx) = \cos(x) + j \cdot \sin(x)$$

FOURIER TRANSFORM

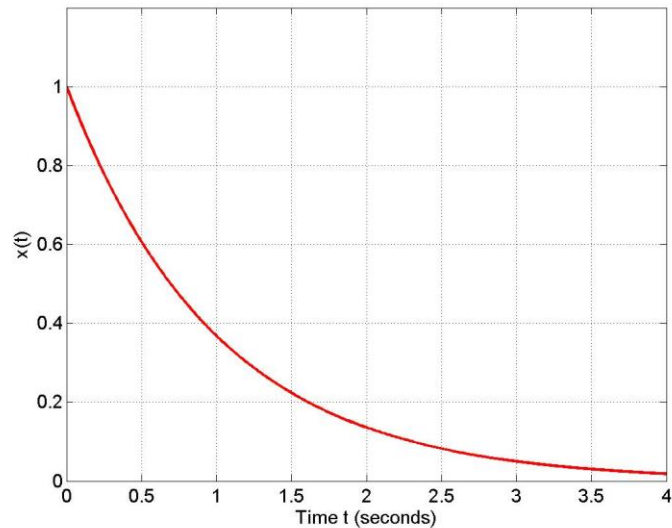
- **Example 1:**

Find the Fourier transform of $g(t) = \exp(-t)$, $t > 0$

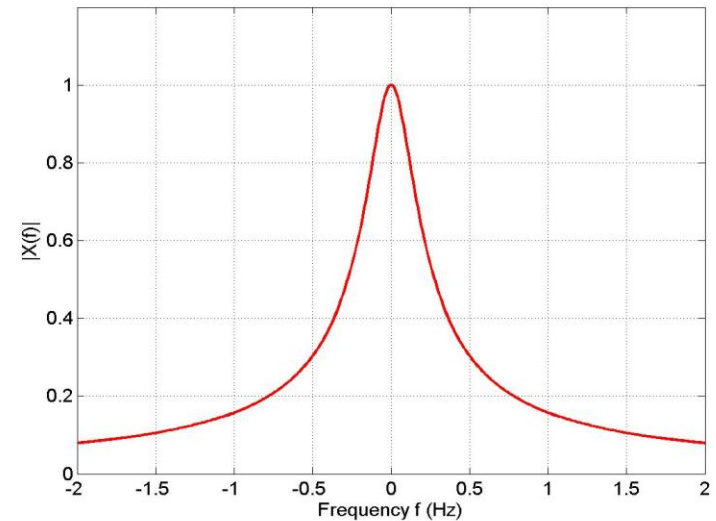
Sol.

FOURIER TRANSFORM

$$s(t) = e^{-t}, \quad t \geq 0 \quad \xrightarrow{\text{Fourier Transform}} \quad S(f) = \frac{1}{1 + j2\pi f}$$



Time Domain



Frequency Domain

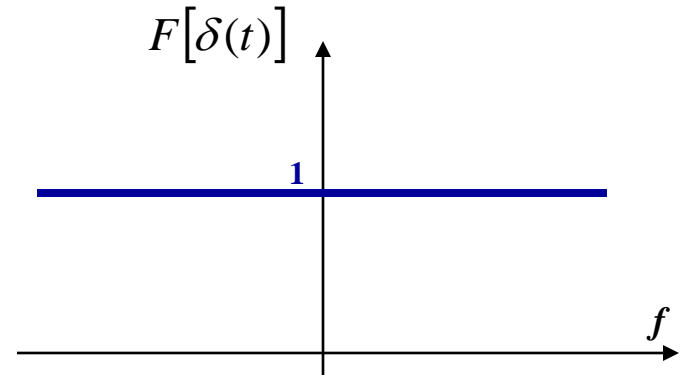
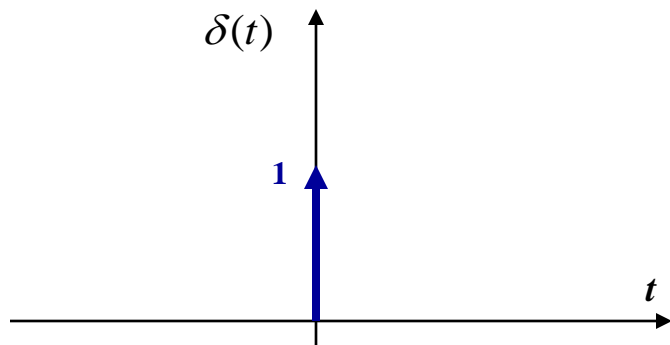
UNIT IMPULSE FUNCTION

- Unit impulse function (delta function)

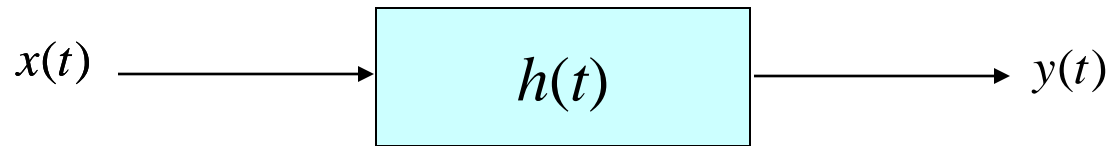
$$\int_{-\infty}^{+\infty} \delta(t) dt = 1, \quad \delta(0) = \infty, \quad \delta(t) = 0, \forall t \neq 0$$

$$\delta(t)x(t) = \delta(t)x(0)$$

$$\int_{-\infty}^{+\infty} \delta(t) \exp(-j2ft) dt =$$



LINEAR TIME INVARIANT (LTI) SYSTEM



- The output of the LTI system is the convolution of input $x(t)$ with the channel impulse response $h(t)$

$$y(t) = x(t) \otimes h(t) = \int_{-\infty}^{+\infty} x(t - \tau)h(\tau)d\tau = \int_{-\infty}^{+\infty} x(\tau)h(t - \tau)d\tau$$

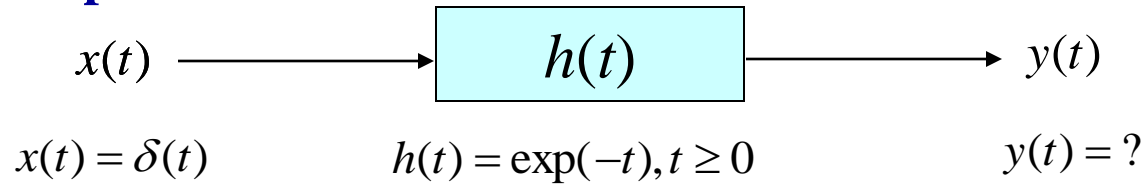
- Time domain convolution \rightarrow frequency domain multiplication

$$X(f) = F[x(t)] \quad H(f) = F[h(t)] \quad Y(f) = F[y(t)]$$

$$Y(f) = X(f) \times H(f)$$

LTI

- **Example:**



Method 1:

Method 2:

CORRELATION

- The cross correlation of two deterministic, complex-valued function $a(t)$ and $b(t)$ is defined as

$$R_{ab}(\tau) = \int_{-\infty}^{+\infty} a(t)b^*(t-\tau)dt$$

- The auto-correlation of one deterministic, complex-valued function $a(t)$ is defined as

$$R_a(\tau) = \int_{-\infty}^{+\infty} a(t)a^*(t-\tau)dt$$

ENERGY

- **The energy of a signal $s(t)$ is defined as**

$$E_s = \int_{-\infty}^{+\infty} |s(t)|^2 dt$$

- If a signal has finite energy, the signal is called energy signal.
- E.g. $s(t) = \exp(-t), t \geq 0$

$$E_s =$$

- **Relationship between energy and auto-correlation function**

$$E_s = \int_{-\infty}^{+\infty} |s(t)|^2 dt = \int_{-\infty}^{+\infty} s(t)s^*(t-0)dt = R_s(0)$$

POWER

- The power of a signal $s(t)$ is defined as

$$P_s = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{+T/2} |s(t)|^2 dt$$

- If a signal has finite power, the signal is called power signal
- For periodic signal with period T

$$P_s = \frac{1}{T} \int_0^T |s(t)|^2 dt$$

- E.g. $s(t) = A \sin(2\pi ft + \theta)$

$$P_s =$$

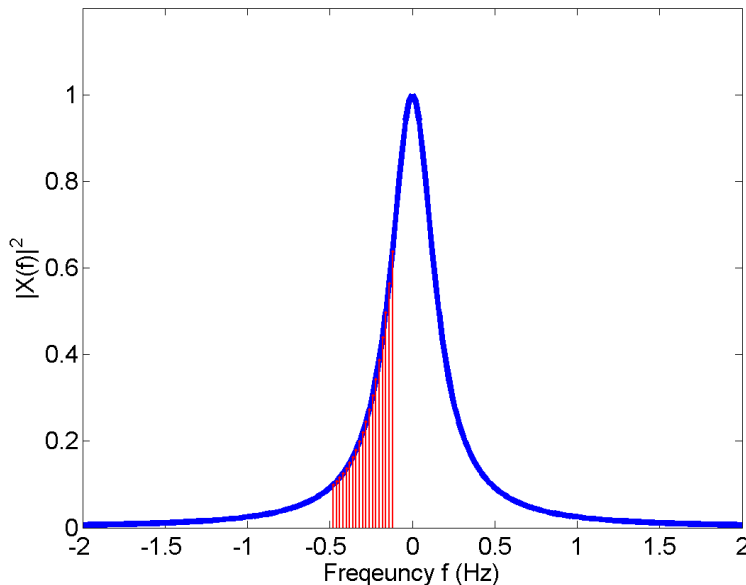
PARSEVAL'S THEOREM & ESD

$$E_s = \int_{-\infty}^{+\infty} |s(t)|^2 dt = \int_{-\infty}^{+\infty} |S(f)|^2 df$$

- $|S(f)|^2$ is called the **energy spectrum density (ESD)** of the signal
 - ESD represents the energy distribution in the frequency domain.

e.g. ESD of $s(t) = \exp(-t), t \geq 0$

$$\frac{1}{|1 + j2\pi f|^2}$$



$\int_{f_1}^{f_2} |S(f)|^2 df$:The signal energy in the frequency range of $[f_1, f_2]$

OUTLINE

- Linear system theory (Fourier analysis) (Appendix A)
- **Random variables and random process (Appendix C)**

DISCRETE RANDOM VARIABLES

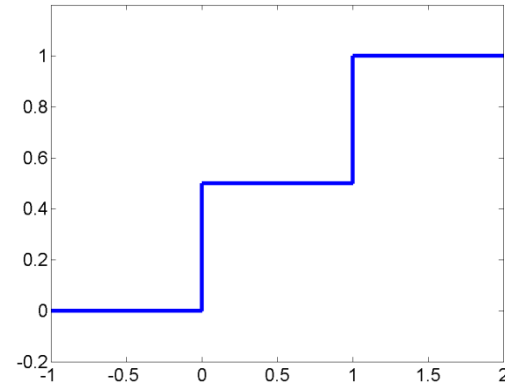
- **Example 1: coin toss**
 - Define random variable (RV) X .
 - $X = 0$: coin head
 - $X = 1$: coin tail
 - **Probability Mass Function (PMF):**
 - $P(X = 0) = 0.5, P(X = 1) = 0.5$
- **Example 2: pick 1 ball from 10 balls (2 black, 3 white, 5 red)**
 - Define RV X
 - $X = 0$: a black ball is picked
 - $X = 1$: a white ball is picked
 - $X = 2$: a red ball is picked
 - PMF:

CUMULATIVE DISTRIBUTION FUNCTION (CDF)

$$F_X(x) = P(X \leq x)$$

- Example 1: toss coin**

$$F_X(x) = \begin{cases} 0, & x < 0 \\ P(X = 0) = 0.5, & 0 \leq x < 1 \\ P(X = 0) + P(X = 1) = 1, & x \geq 1 \end{cases}$$



- Example 2: pick 1 balls from 10 balls**

$$F_X(x) = \begin{cases} x < 0 \\ 0 \leq x < 1 \\ 1 \leq x < 2 \\ x \geq 2 \end{cases}$$

CONTINUOUS RV

- **RV X can take continuous values**
 - E.g.: X represents the average temperature

- **CDF**

$$F_X(x) = P(X \leq x)$$

$$F_X(-\infty) = 0 \quad F_X(\infty) = 1$$

- **Probability density function (pdf)**

$$f_X(x) = \frac{dF_X(x)}{dx}$$

Interpretation: for small Δx ,

$$f_X(x) = \frac{F_X(x + \Delta x) - F_X(x)}{\Delta x} = \frac{P(x < X < x + \Delta x)}{\Delta x}$$

$$f_X(x)\Delta x = P(x < X < x + \Delta x)$$

or, the probability that the RV X is in the range of $(x, x + \Delta x)$

PDF AND CDF

- Relationship between pdf and CDF

$$f_X(x) = \frac{dF_X(x)}{dx}$$

$$F_X(x) = \int_{-\infty}^x f_X(y)dy$$

$$P(x_2 < X < x_1) = F_X(x_2) - F_X(x_1) = \int_{x_1}^{x_2} f_X(y)dy$$

$$P(X < \infty) = F_X(\infty) = \int_{-\infty}^{+\infty} f_X(y)dy = 1$$

CDF AND PDF

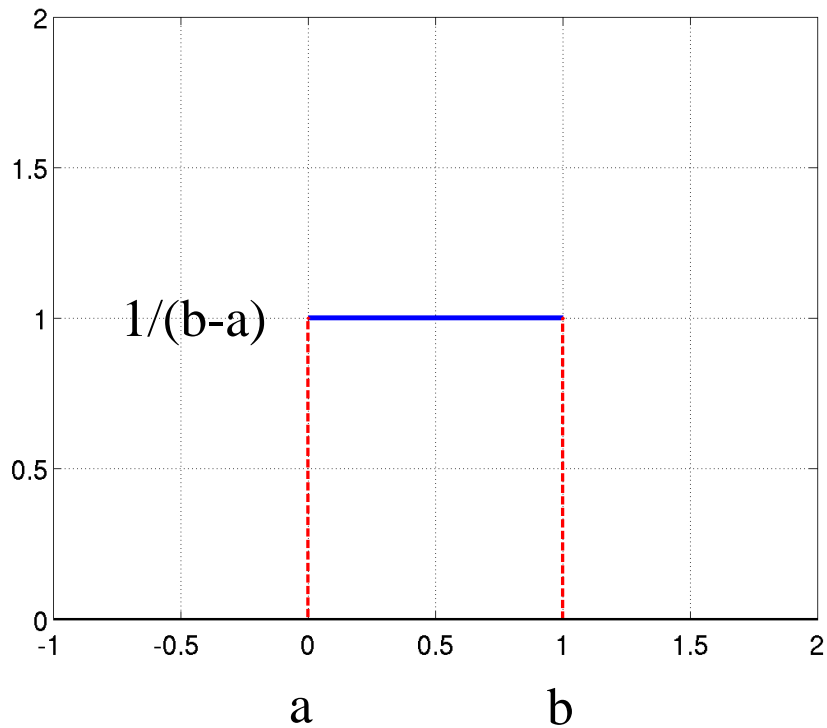
- **Example:**

The pdf of uniform distribution is: $f_X(x) = \frac{1}{b-a}, \quad a < x < b$

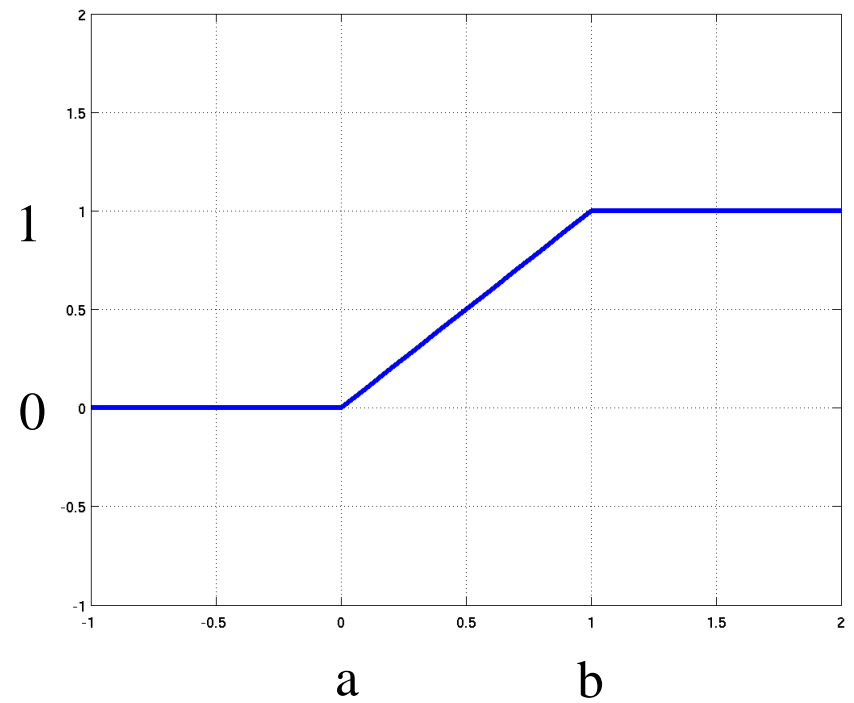
The corresponding CDF is:

Let $a = 0, b = 1$, the probability that X is in the range of $[0.5, 0.6]$ is:

CDF AND PDF



pdf of uniform distribution



cdf of uniform distribution

EXPECTATION (MEAN)

- **Discrete-time RV**

$$m_X = E(X) = \sum_i x_i P(X = x_i)$$

- Weighted sum

- **Continuous-time RV**

$$m_X = E(X) = \int_{-\infty}^{+\infty} y f_X(y) dy$$

- **Example: find the expectation (mean) of the uniform distribution**

$$f_X(x) = \frac{1}{b-a}, \quad a < x < b$$

VARIANCE

- **Discrete RV**

$$\sigma^2 = \mathbb{E}[(X - m_X)^2] = \sum_i (x_i - m_X)^2 P(X = x_i)$$

- **Continuous RV**

$$\sigma^2 = \mathbb{E}[(X - m_X)^2] = \int_{-\infty}^{+\infty} (y - m_X)^2 f_X(y) dy$$

-

$$\sigma^2 = \mathbb{E}(X^2) - m_X^2$$

– Proof:

VARIANCE

- **Example: find the variance of the uniform distribution**

$$f_X(x) = \frac{1}{b-a}, \quad a < x < b$$

– Sol:

GAUSSIAN DISTRIBUTION

- The pdf of Gaussian distribution is

$$f_X(x) = \frac{1}{\sqrt{2\pi}\sigma} \exp\left[-\frac{(x-m)^2}{2\sigma^2}\right]$$

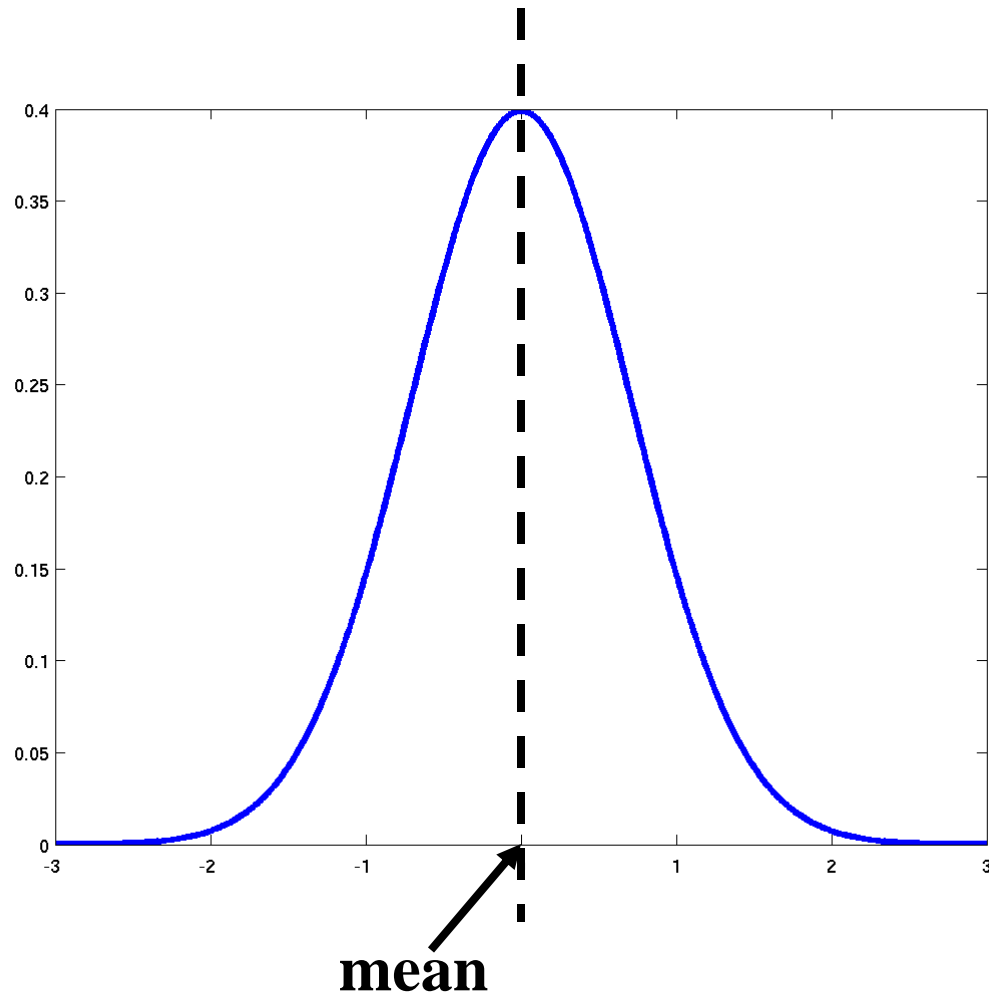
- The pdf is determined by two parameters:

$$m = E(X) \quad \text{mean}$$

$$\sigma = \sqrt{E(X - m_X)^2} \quad \text{standard deviation}$$

- The sum of Gaussian RVs is still Gaussian distributed

GAUSSIAN DISTRIBUTION



JOINT DISTRIBUTION

- Consider two RVs, X , and, Y . The joint CDF of X and Y is

$$F_{X,Y}(x, y) = P(X \leq x, Y \leq y)$$

- Joint pdf

$$f_{X,Y}(x, y) = \frac{\partial F_{X,Y}(x, y)}{\partial x \partial y}$$

- Given the joint pdf, we can find the marginal pdf of each RV

$$f_X(x) = \int_{-\infty}^{+\infty} f_{X,Y}(x, y) dy \qquad f_Y(y) = \int_{-\infty}^{+\infty} f_{X,Y}(x, y) dx$$

JOINT DISTRIBUTION

- **Example:**

$$f_{X,Y}(x, y) = \begin{cases} x + y, & 0 \leq x \leq 1, 0 \leq y \leq 1 \\ 0, & \text{otherwise} \end{cases}$$

$$f_X(x) = ?, f_Y(y) = ?$$

– Sol:

- **Independent**

– If $f_{X,Y}(x, y) = f_X(x)f_Y(y)$, then X and Y are independent.

CORRELATION

- The correlation of two RVs X and Y is calculated as

$$E[XY] = \int_{-\infty}^{+\infty} \int_{-\infty}^{+\infty} xy f_{XY}(x, y) dx dy$$

- The covariance of two RVs X and Y is

$$E[(X - m_X)(Y - m_Y)] = E[x]E[Y] - m_X m_Y$$

- If $E[XY] = E(X)E(Y)$, then X and Y are uncorrelated.

If X and Y are independent \rightarrow they must be uncorrelated.

Proof:

CORRELATION

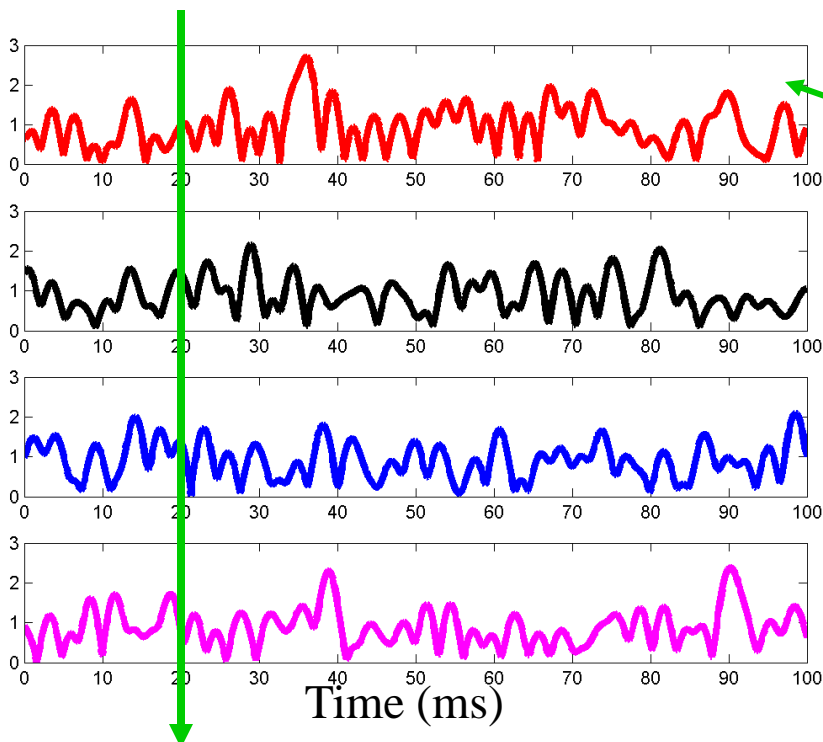
- **Example:**
$$f_{X,Y}(x, y) = \begin{cases} x + y, & 0 \leq x \leq 1, 0 \leq y \leq 1 \\ 0, & \text{otherwise} \end{cases}$$

Are X, Y uncorrelated?

– Sol:

RANDOM PROCESS

- **Random process $X(t)$: an RV changes w.r.t. time**
 - $X(t)$ is a function of time t
 - At any time instant t_0 , $X(t_0)$ is a random variable.



Each realization is called a **sample function** of the random process

mean: (ensemble average)

$$m_X(t_0) = E[X(t_0)]$$

At each time instant, we have a RV

RANDOM PROCESS

- Let $X(t)$ be a random process,
at time t_1 , we have a RV $X(t_1)$;
at time t_2 , we have another RV $X(t_2)$.

Then we have the joint distribution functions

$$F_{X(t_1)X(t_2)}(x_1, x_2) = P[X(t_1) < x_1, X(t_2) < x_2]$$

$$f_{X(t_1)X(t_2)}(x_1, x_2) = \frac{\partial F_{X(t_1)X(t_2)}(x_1, x_2)}{\partial x_1 \partial x_2}$$

AUTO-CORRECTION FUNCTION (ACF)

- Let $X(t)$ be a random process,
at time t_1 , we have a RV $X(t_1)$;
at time t_2 , we have another RV $X(t_2)$.

Then we can calculate the correlation of $X(t_1)$ and $X(t_2)$:

$$R_X(t_1, t_2) = E[X(t_1)X^*(t_2)]$$

$R_X(t_1, t_2)$ **is defined as the autocorrelation function of $X(t)$**

WIDE-SENSE STATIONARY (WSS)

- A random process is wide-sense stationary (WSS) if the following two conditions are satisfied

1. $m_X(t) = E[X(t)] = m_x$

2. $R_X(t_1, t_2) = R_X(t_1 + h, t_2 + h) = R_X(t_1 - t_2)$

- The first condition states that the mean of the random process is independent of time.
- The second condition states that the auto-correlation function is only dependent on the time difference between the two RVs, and it's independent of the starting time.

$$E[X(t_1)X(t_2)] = E[X(t_1 + h)X(t_2 + h)]$$

- The autocorrelation function of WSS process is usually represented as

$$R_X(\tau)$$

POWER SPECTRUM DENSITY

- The power spectrum density (PSD) of WSS random process is defined as the Fourier transform of the auto-correlation function of the process

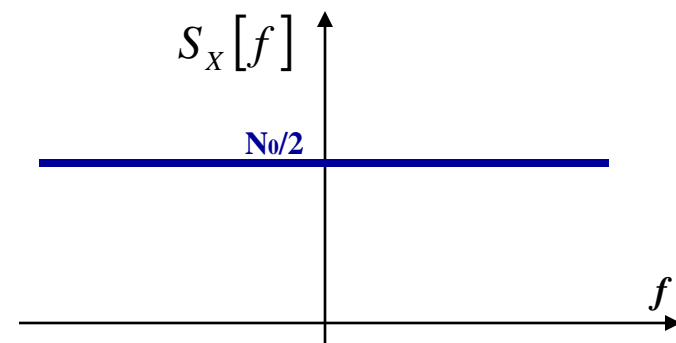
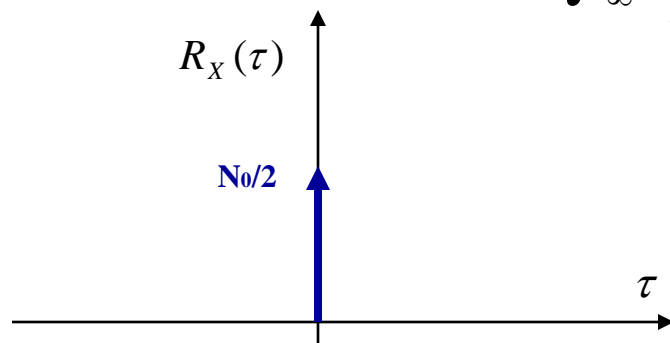
$$S_X(f) = \int_{-\infty}^{+\infty} R_X(\tau) \exp(-j2\pi f\tau) d\tau$$

- Power spectrum density represents the power distribution in the frequency domain.

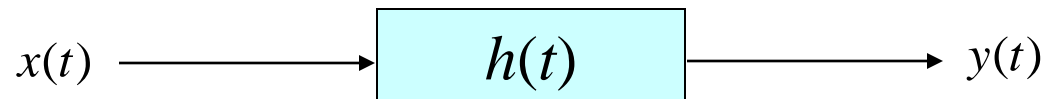
- E.g.

$$R_x(\tau) = \frac{N_0}{2} \delta(\tau)$$

$$S_x(f) = \int_{-\infty}^{+\infty} \frac{N_0}{2} \delta(\tau) d\tau = \frac{N_0}{2}$$



RANDOM PROCESS PASS THROUGH LTI



- If $x(t)$ is a WSS random process, then $y(t)$ is a WSS random process as well.

The relationship between mean of $y(t)$ and mean of $x(t)$

$$E[y(t)] = E\left[\int_{-\infty}^{+\infty} h(t-\tau)x(\tau)d\tau\right] = \int_{-\infty}^{+\infty} h(t-\tau)E[x(\tau)]d\tau$$

The relationship between PSD of $y(t)$ and PSD of $x(t)$

$$S_Y(f) = |H(f)|^2 S_X(f)$$